Independent Component Analysis

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This lecture's overview

- A motivating example
- The theory
 - Decorrelation
 - Independence vs decorrelation
 - Independent component analysis
- Separating sounds
 - Solving instantaneous mixtures
 - Solving convolutive mixtures
- Data exploration and independence
 - Extracting audio features
 - Extracting multimodal features

A "simple" audio problem



Formalizing the problem

- Each mic will receive a mix of both sounds
 - Sound waves superimpose linearly
 - We'll ignore propagation delays for now
- The simplified mixing model is:

$$\mathbf{x}(t) = \mathbf{A} \cdot \mathbf{s}(t)$$

We know x(t), but nothing else
How do we solve this system and find s(t)?



 $x_2(t) = a_{21}s_1(t) + a_{22}s_2(t)$

When can we solve this?

- The mixing equation is: $\mathbf{x}(t) = \mathbf{A} \cdot \mathbf{s}(t)$
- Our estimates of $\mathbf{s}(t)$ will be: $\hat{\mathbf{s}}(t) = \mathbf{A}^{-1} \cdot \mathbf{x}(t)$
- To recover s(*t*), A must be invertible:
 - We need as many mics as sources
 - The mics/sources must not coincide
 - All sources must be audible
- Otherwise this is a different story ...





A simple example



- A simple invertible problem
 - **s**(*t*) contains two structured waveforms
 - A is invertible (but we don't know it)
 - **x**(*t*) looks messy, doesn't reveal **s**(*t*) clearly
- How do we solve this? Any ideas?

What to look for

$$\mathbf{x}(t) = \mathbf{A} \cdot \mathbf{s}(t)$$

- We can only use **x**(*t*)
- Is there a property we can take advantage of?
- Yes! We know that different sounds are "statistically unrelated"
- The plan: Find a solution that enforces this "unrelatedness"

A first try

- Find s(t) by minimizing cross-correlation
- Our estimate of **s**(*t*) is computed by:

$$\hat{\mathbf{s}}(t) = \mathbf{W} \cdot \mathbf{x}(t)$$

- If $\mathbf{W}\approx\mathbf{A}^{\text{-1}}$ then we have a good solution
- The goal is that the output becomes uncorrelated:

$$\langle \hat{s}_i(t) \cdot \hat{s}_j(t) \rangle = 0, \forall i \neq j$$

- · We assume here that our signals are zero mean
- So the overall problem to solve is:

$$\underset{\mathbf{w}}{\operatorname{arg\,min}}\left\langle \sum_{k} a_{ik} x_{k}(t) \cdot \sum_{k} a_{jk} x_{k}(t) \right\rangle, \forall i \neq j$$

.

How to solve for uncorrelatedness

- Let's use matrices instead of time-series:

$$\mathbf{x}(t) \to \mathbf{X} = \begin{bmatrix} x_1(1), & \cdots, & x_1(N) \\ x_2(1), & \cdots, & x_2(N) \end{bmatrix}, etc \qquad \mathbf{x}(t) = \mathbf{A} \cdot \mathbf{s}(t) \text{ and } \hat{\mathbf{s}}(t) = \mathbf{W} \cdot \mathbf{x}(t) \Rightarrow$$
$$\Rightarrow \mathbf{X} = \mathbf{A} \cdot \mathbf{S} \text{ and } \hat{\mathbf{S}} = \mathbf{W} \cdot \mathbf{X}$$

• The uncorrelatedness translates to:

$$\frac{1}{N}\hat{\mathbf{S}}\cdot\hat{\mathbf{S}}^{T} = \begin{bmatrix} c_{11} & 0\\ 0 & c_{22} \end{bmatrix}$$

• We then need to diagonalize:

$$\frac{1}{N}\hat{\mathbf{S}}\cdot\hat{\mathbf{S}}^{T} = \frac{1}{N}(\mathbf{W}\cdot\mathbf{X})(\mathbf{W}\cdot\mathbf{X})^{T}$$

How to solve for uncorrelatedness

- This is actually a well known problem in linear algebra
- One solution is Eigenanalysis:

$$Cov(\hat{\mathbf{S}}) = \frac{1}{N} \hat{\mathbf{S}} \cdot \hat{\mathbf{S}}^{T}$$
$$= \frac{1}{N} (\mathbf{W} \cdot \mathbf{X}) (\mathbf{W} \cdot \mathbf{X})^{T}$$
$$= \frac{1}{N} \mathbf{W} \cdot \mathbf{X} \cdot \mathbf{X}^{T} \cdot \mathbf{W}^{T}$$
$$= \mathbf{W} \cdot Cov(\mathbf{X}) \cdot \mathbf{W}^{T}$$

Cov(X) is a symmetric matrix

How to solve for uncorrelatedness

• For any symmetric matrix **Z** we know that:

$$\mathbf{Z} = \mathbf{U} \cdot \mathbf{D} \cdot \mathbf{U}^{T} = \begin{bmatrix} \uparrow & \uparrow \\ \mathbf{u}_{1} & \cdots & \mathbf{u}_{N} \\ \downarrow & \downarrow \end{bmatrix} \cdot \begin{bmatrix} d_{1} & & \\ & \ddots & \\ & & d_{N} \end{bmatrix} \cdot \begin{bmatrix} \leftarrow & \mathbf{u}_{1} & \rightarrow \\ \leftarrow & \mathbf{u}_{N} & \rightarrow \end{bmatrix}$$

- Where **u**_i and d_i are **Z**'s eigenvectors and eigenvalues respectively
- In our case if [U,D] = eig(Cov(X))

$$Cov(\hat{\mathbf{S}}) = \mathbf{W} \cdot Cov(\mathbf{X}) \cdot \mathbf{W}^{T}$$

= $\mathbf{W} \cdot \mathbf{U} \cdot \mathbf{D} \cdot \mathbf{U}^{T} \cdot \mathbf{W}^{T}$ let us replace \mathbf{W} with \mathbf{U}^{T}
= $\mathbf{U}^{T} \cdot \mathbf{U} \cdot \mathbf{D} \cdot \mathbf{U}^{T} \cdot \mathbf{U}$ U is orthonormal $(\mathbf{U} \cdot \mathbf{U}^{T} = \mathbf{I})$
= \mathbf{D}

- This is actually Principal Component Analysis

Another solution

• We can also solve for a matrix inverse square root:

$$Cov(\hat{\mathbf{S}}) \propto \mathbf{W} \cdot \mathbf{X} \cdot \mathbf{X}^{T} \cdot \mathbf{W}^{T}$$

$$= (\mathbf{X} \cdot \mathbf{X}^{T})^{-\frac{1}{2}} \cdot \mathbf{X} \cdot \mathbf{X}^{T} \cdot (\mathbf{X}^{T} \cdot \mathbf{X})^{-\frac{1}{2}} \quad \text{replace } \mathbf{W} \text{ with } (\mathbf{X} \cdot \mathbf{X}^{T})^{-\frac{1}{2}}$$

$$= \mathbf{I}$$

$$(\mathbf{X} \cdot \mathbf{X})^{-\frac{1}{2}} = \mathbf{U} \cdot \begin{bmatrix} d_{1}^{-\frac{1}{2}} & & \\ & \ddots & \\ & & d_{N}^{-\frac{1}{2}} \end{bmatrix} \cdot \mathbf{U}^{T}, \quad \text{where } [\mathbf{U}, \mathbf{D}] = eig(\mathbf{X} \cdot \mathbf{X}^{T})$$

Another approach

- What if we want to do this in real-time?
- We can also estimate W in an online manner:

$$\Delta \mathbf{W} \propto \mu \left(\mathbf{I} - \mathbf{W} \cdot \mathbf{x}(t) \cdot \mathbf{x}(t)^T \cdot \mathbf{W}^T \right) \mathbf{W}$$

- Every time we see a new observation $\mathbf{x}(t)$ we update \mathbf{W}
- Using this adaptive approach we can see that: $Cov(\mathbf{W} \cdot \mathbf{x}(t)) = \mathbf{I}, \quad \Delta \mathbf{W} = 0$
- · We'll skip the derivation details for now

Summary so far

• For a mixture:

 $\mathbf{X} = \mathbf{A} \cdot \mathbf{S}$

• We can algebraically recover an uncorrelated output using

 $\hat{\mathbf{S}} = \mathbf{W} \cdot \mathbf{X}$

- If W is the eigenvector matrix of $\mathit{Cov}(\mathbf{X})$
- Or with $\mathbf{W} = Cov(\mathbf{X})^{-1/2}$
- Or we can use an online estimator:

 $\Delta \mathbf{W} \propto \mu \left(\mathbf{I} - \mathbf{W} \cdot \mathbf{x}(t) \cdot \mathbf{x}(t)^T \cdot \mathbf{W}^T \right) \mathbf{W}$



• Well, that was a waste of time ...

What went wrong?

- What does decorrelation mean?
 - That the two things compared are "not related"
- Consider a mixture of two Gaussians



Decorrelation

• Now let us do what we derived so far on this signal:



- After we are done the two Gaussians are "statistically independent"
 i.e., P(s₁,s₂) = P(s₁)P(s₂)
- We have in effect separated the original signals
 Save for a scaling ambiguity



Doing PCA doesn't give us the right solution

- The result is not what we want
 - We are off by a rotation
- This idea doesn't seem to work for non-Gaussian signals

a) Find the eigenvectors

b) Rotate and scale so that covariance is I



So what's wrong?

- · For Gaussian data decorrelation means independence
 - Gaussians have up to second order statistics (1st is mean, 2nd is variance)
 - By minimizing the 2nd-order cross-statistics we achieve independence
 - These statistics can be expressed by the 2nd-order cumulants: $cum(x_i,x_i) = \left< x_i x_i \right>$
 - · Which happen to be the diagonals of the covariance matrix

- But real-world data are seldom Gaussian

- · Non-Gaussian data have higher orders which are not taken care of with PCA
- We can measure their dependence using higher order cumulants: $\begin{array}{l}3^{\prime\prime} \ order: \ \ cum(x_i,x_j,x_k) = \left\langle x_i x_j x_k \right\rangle \\ 4^{\ast} \ \ order: \ \ cum(x_i,x_j,x_k,x_i) = \left\langle x_i x_j x_k x_i \right\rangle - \left\langle x_i x_j \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle \left\langle x_i x_i \right\rangle - \left\langle x_i x_i \right\rangle -$



The real problem to solve

- For statistical independence we need to minimize <u>all</u> cross-cumulants
 In practice up to 4th order is enough
- For 2nd order we minimized the off-diagonal covariance elements

 $\begin{bmatrix} cum(x_1,x_1) & cum(x_1,x_2) \\ cum(x_2,x_1) & cum(x_2,x_2) \end{bmatrix}$

- For 4th order we will do the same for a tensor

 $Q_{i,i,k,l} = cum(x_i, x_i, x_k, x_l)$

- The process is similar to PCA, but in more dimensions
 We now find *"eigenmatrices"* instead of eigenvectors
- Algorithms like JADE and FOBI solve this problem
 - Can you see a potential problem though?

An alternative approach

- Tensorial methods can be very very computationally intensive
- How about an on-line method instead?
- Independence can also be coined as "non-linear decorrelation"
 - x and y are independent if and only if:

$\langle f(x)g(y)\rangle = \langle f(x)\rangle\langle g(y)\rangle$

- For all continuous functions *f* and *g*
- This is a non-linear extension of 2^{nd} order independence where f(x) = g(x) = x
- We can try solving for that then

Online ICA

- Conceptually this is very similar to online decorrelation
 - For decorrelation:
 - $\Delta \mathbf{W} \propto \mu \left(\mathbf{I} \mathbf{W} \cdot \mathbf{x}(t) \cdot \mathbf{x}(t)^T \cdot \mathbf{W}^T \right) \mathbf{W}$
 - · For non-linear decorrelation:

$$\Delta \mathbf{W} \propto \mu \left(\mathbf{I} - f \left(\mathbf{W} \cdot \mathbf{x}(t) \right) \cdot g \left(\mathbf{W} \cdot \mathbf{x}(t) \right)^T \right) \mathbf{W}$$

- This adaptation method is known as the Cichocki-Unbehauen update
 - But we can obtain it using many different ways
- But how do we pick the non-linearities?
 - · Depends on the prior we have on the sources

$$f(x_i) = \begin{cases} x + \tanh(x), \text{ for super-Gaussians} \\ x - \tanh(x), \text{ for sub-Gaussians} \end{cases}$$

Other popular approaches

- Minimum Mutual Information
 - Minimize the mutual information of the output
 - · Creates maximally statistically independent outputs
- Infomax
 - Maximize the entropy of the output or Mutual Information of input/output
- Non-Gaussianity
 - Adding signals tends towards Gaussianity (Central Limit Theorem)
 - · Find the maximally non-Gaussian outputs undoes the mixing
- Maximum Likelihood
 - · Less straightforward at first, but elegant nevertheless
- Geometric methods
 - Trying to "eyeball" the proper way to rotate





• We actually separated the mixture!





Problems with instantaneous mixing

- Sounds don't really mix instantaneously
- There are multiple effects
 - Room reflections
 - Sensor response
 - Propagation delays
 - Propagation and reflection filtering
- Most can be seen as filters
- We need a *convolutive mixing* model







Convolutive mixing

Instead of instantaneous mixing:

$$x_i(t) = \sum_{j=1}^{n} a_{ij} s_j(t)$$

• We now have *convolutive* mixing:

$$x_i(t) = \sum_j \sum_k a_{ij}(k) s_j(t-k)$$

- The mixing filters a_{ij}(k) encapsulate all the mixing effects in this model
- But how do we do ICA now?
 - This is an ugly equation!





FIR matrix algebra

- Matrices with FIR filters as elements

$$\underline{\mathbf{A}} = \begin{bmatrix} \underline{a}_{11} & \underline{a}_{12} \\ \underline{a}_{21} & \underline{a}_{22} \end{bmatrix}$$
$$\underline{a}_{ij} = \begin{bmatrix} a_{ij}(0) & \cdots & a_{ij}(k-1) \end{bmatrix}$$

• FIR matrix multiplication performs convolution and accumulation

$$\underline{\mathbf{A}} \cdot \underline{\mathbf{b}} = \begin{bmatrix} \underline{a}_{11} & \underline{a}_{12} \\ \underline{a}_{21} & \underline{a}_{22} \end{bmatrix} \cdot \begin{bmatrix} \underline{b}_1 \\ \underline{b}_2 \end{bmatrix} = \begin{bmatrix} \underline{a}_{11} * \underline{b}_1 + \underline{a}_{12} * \underline{b}_2 \\ \underline{a}_{21} * \underline{b}_1 + \underline{a}_{22} * \underline{b}_2 \end{bmatrix}$$

Back to convolutive mixing

 Now we can rewrite convolutive mixing as:

$$\begin{aligned} x_i(t) &= \sum_j \sum_k a_{ij}(k) s_j(t-k) \Rightarrow \\ \Rightarrow \mathbf{\underline{x}}(t) &= \mathbf{\underline{A}} \cdot \mathbf{\underline{s}}(t) = \begin{bmatrix} a_{11} * s_1(t) + a_{12} * s_2(t) \\ a_{21} * s_1(t) + a_{22} * s_2(t) \end{bmatrix} \end{aligned}$$

- Tidier formulation!
- We can use the FIR matrix abstraction to solve this problem now



An easy way to solve convolutive mixing

• Straightforward translation of instantaneous learning rules using FIR matrices:

$$\Delta \underline{\mathbf{W}} \propto \left(\mathbf{I} + f(\underline{\mathbf{W}} \cdot \underline{\mathbf{x}}) \cdot (\underline{\mathbf{W}} \cdot \underline{\mathbf{x}})^T \right) \cdot \underline{\mathbf{W}}$$

- Not so easy with algebraic approaches!
- Multiple other (and more rigorous/better behaved) approaches have been developed

Complications with this approach

- Required convolutions are expensive
 - Real-room filters are long
 - Their FIR inverses are very long
 - FIR products can become very time consuming
- Convergence is hard to achieve
 - Huge parameter space
 - Tightly interwoven parameter relationships
- A slow optimization nightmare!

FIR matrix algebra, part II

• FIR matrices have frequency domain counterparts:

$$\underline{\mathbf{A}} = \begin{bmatrix} \underline{a}_{11} & \underline{a}_{12} \\ \underline{a}_{21} & \underline{a}_{22} \end{bmatrix} \xrightarrow{\text{frequency domain}} \hat{\mathbf{A}} = \begin{bmatrix} \underline{\hat{a}}_{11} & \underline{\hat{a}}_{12} \\ \underline{\hat{a}}_{21} & \underline{\hat{a}}_{22} \\ \underline{\hat{a}}_{ij} = \text{DFT}[\underline{a}_{ij}]$$

- And their products are simpler:

$$\hat{\underline{\mathbf{A}}} \cdot \hat{\underline{\mathbf{b}}} = \begin{bmatrix} \hat{a}_{11} \cdot \hat{\underline{b}}_1 + \hat{a}_{12} \cdot \hat{\underline{b}}_2 \\ \hat{\underline{a}}_{21} \cdot \hat{\underline{b}}_1 + \hat{\underline{a}}_{22} \cdot \hat{\underline{b}}_2 \end{bmatrix}$$
$$\hat{\underline{a}} \cdot \hat{\underline{b}} = \begin{bmatrix} a(0) \cdot b(0) & \cdots & a(k-1) \cdot b(k-1) \end{bmatrix}$$



- We can now model the process in the frequency domain: $\hat{\mathbf{X}} = \hat{\mathbf{A}} \cdot \hat{\mathbf{S}}$
- For every frequency we have: $\mathbf{X}_{f}(t) = \mathbf{A}_{f} \cdot \mathbf{S}_{f}(t), \ \forall f, t$
- Hey, that's instantaneous mixing!
 - We can solve that!





Some complications ...

- Permutation issues
 - · We don't know which source will end up in each narrowband output ...
 - Resulting output can have separated narrowband elements from both sounds!



- Scaling issues
 - Narrowband outputs can be scaled arbitrarily
 - This results in spectrally colored outputs





Scaling issue

One simple fix is to normalize the separating matrices

$$\mathbf{W}_{f}^{norm} = \mathbf{W}_{f}^{orig} \cdot \left| \mathbf{W}_{f}^{orig} \right|^{\frac{1}{N}}$$

- Results into more reasonable scaling
- More sophisticated approaches exist but this is not a major problem
- Some spectral coloration is however unavoidable



Some solutions for permutation problems

- Continuity of unmixing matrices
 - Adjacent unmixing matrices tend to be a little similar, we can permute/bias them accordingly
 - Doesn't work that great
- Smoothness of spectral output
 - Narrowband components from each source tend to modulate the same way
 - Permute unmixing matrices to ensure adjacent narrowband output are similarly modulated
 - Works fine
- The above can fail miserably for more than two sources!
 - Combinatorial explosion!

Beamforming and ICA

- If we know the placement of the sensors we can obtain the spatial response of the ICA solution
- ICA places nulls to cancel out interfering sources
 - Just as in the instantaneous case we cancel out sources
- We can visualize the permutation problem now
 - Out of place bands



Corrected source



Using beamforming to resolve permutations

- Spatial information can be used to resolve permutations
 - Find permutations that preserve zeros or smooth out the responses
- Works fine, although it can be flaky if the array response is not that clean







The N-input N-output problem

- ICA, in either formulation inverts a square matrix (whether scalar, or FIR)
- This implies that we have the same number of input as outputs
- E.g. in a street with 30 noise sources we need at least 30 mics!
- Solutions exist for *M* ins *N* outs where *M* > *N*
- If *N* > *M* we can only beamform
 - In some cases extra sources can be treated as noise
- · This can be restrictive in some situations

Separation recap

- Orthogonality is not independence!!
 - Not all signals are Gaussian which is a usual assumption
- · We can model instantaneous mixtures with ICA and get good results
 - ICA algorithms can optimize a variety of objectives, but ultimately result in statistical independence between the outputs
 - · Same model is useful for all sorts of mixing situations
- Convolutive mixtures are more challenging but solvable
 - · There's more ambiguity, and a closer link to signal processing approaches

ICA for data exploration

- ICA is also great for data exploration
- If PCA is, then ICA should be, right?
- With data of large dimensionalities we want to find structure
- PCA can reduce the dimensionality
 And clean up the data structure a bit
- But ICA can find much more intuitive projections



Example cases of PCA vs ICA

- Motivation for using ICA vs PCA
- PCA will indicate orthogonal directions of maximal variance
 - This is great for Gaussian data
 - Also great if we are into LS models
- Real-world is not Gaussian though
- ICA finds directions that are more "revealing"

Non-Gaussian data



Finding useful transforms with ICA

- Audio preprocessing example
- Take a lot of audio snippets and concatenate them in a big matrix, do component analysis
- PCA results in the DCT bases
 Do you see why?
- ICA returns time/freq localize sinusoids which is a better way to analyze sounds
- Ditto for images
 - ICA returns localizes edge filters

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Enhancing PCA with ICA

- ICA cannot perform dimensionality reduction
 - · The goal is to find independent components, hence there is no sense of order
- · PCA does a great job at reducing dimensionality
 - · Keeps the elements that carry most of the input's energy
- It turns out that PCA is a great preprocessor for ICA
 - There is no guarantee that the PCA subspace will be appropriate for the independent
 components but for most practical purposes this doesn't make a big difference





A Video Example

- The movie is a series of frames
 - · Each frame is a data point
 - 126, 80×60 pixel frames
 - Data X will be 4800×126
- Using PCA/ICA
 - $\mathbf{X} = \mathbf{W} \times \mathbf{H}$
 - W will contain visual components
 - H will contain their time weights





PCA Results

- Nothing special about the visual components
- They are orthogonal pictures
- Does this mean anything? (not really ...)
- Some segmentation of constant vs. moving parts
- Some highlighting of the action in the weights



	Component weights
1	
2	~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~
3	
	T .

ICA Results

- Much more interesting visual components
- They are *independent*
 - Unrelated elements (left/ right hands, background) are now highlighted
 - We have some decomposition by parts
- Components weights are now describing the scene



	Component weights
1	
	\neg \neg \neg \neg
5	
3	Time

A Video Example

- The movie is a series of frames
 - Each frame is a data point
 - 315, 80×60 pixel frames
 - Data X will be 4800×315
- Using PCA/ICA
 - $\mathbf{X} = \mathbf{W} \times \mathbf{H}$
 - W will contain visual components
 - H will contain their time weights





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What about the soundtrack?

- We can also analyze audio in a similar way
- We do a frequency transform and get an audio spectrogram X
 - X is frequencies × time
 - Distinct audio elements can be seen in X
- Unlike before we have only one input this time





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PCA on Audio

• Umm ... it sucks!

useless

 Orthogonality doesn't mean much for audio components

Results are mathematically

optimal, perceptually

200			WARD								
150											
100											
50											
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Input

- ICA on Audio
- A definite improvement
- Independence helps pick up somewhat more meaningful sound objects
- Not too clean results, but the intentions are clear
- Misses some details





Audio Visual Components?

- We can can even take in both audio and video data and try to find structure
- Sometimes there is a very strong correlation between auditory and visual elements
- We should be able to discover that automatically





Audio/Visual Components



Component weights



Which allows us to play with output

 And of course once we have such a nice description we can resynthesize at will



Recap

- A motivating example
- The theory
 - Decorrelation
 - Independence vs decorrelation
 - Independent component analysis
- Separating sounds
 - Solving instantaneous mixtures
 - Solving convolutive mixtures
- Data exploration and independence
 - Extracting audio features
 - Extracting multimodal features

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